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Paper Authors

***T. SRAVAN KUMAR , SRI. R. SATHISH KUMAR.**

* Dept of EEE, QIS College of Engineering& Technology.



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HYBRID FUZZY CONTROLLED SOLAR POWERED SRM FOR WATER PUMPING APPLICATION

¹T. SRAVAN KUMAR, ²SRI. R. SATHISH KUMAR

*PG Scholar, Dept of EEE, QIS College of Engineering & Technology, Ongole, A.P, India,

**Associate Professor, Dept of EEE, QIS College of Engineering & Technology, Ongole, A.P, India

ABSTRACT:

This project presents solar powered SRM with fuzzy controller for water pumping application. Among all special machines, the switched reluctance motor (SRM) has the simplest construction. The rotor of SRM contains no conductors or permanent magnets. Because of its high torque density, low inertia, quick response, variable losses and wide speed range capability, SRM could be a suitable candidate for water pumping powered by SPV array. The speed of SRM is controlled by varying the DC-bus voltage of the mid-point converter. A DC-DC Cuk converter operating in continuous conduction mode (CCM) is used for DC-bus voltage control. The adjustment in step size of an incremental conductance MPPT algorithm facilitates the soft starting of SRM drive. The PV array converts the received solar energy to electrical energy, which is fed to the switched reluctance motor. Further to add to its features minimal rule based hybrid fuzzy logic speed controller is introduced. The performance characteristics of the SRM drive system with fuzzy logic controller are improved and it is observed by using MATLAB/SIMULINK software.

Index Terms: Continuous conduction mode (CCM), Cuk converter, Incremental conductance (InC), Maximum power point tracking (MPPT), Switched reluctance motor (SRM), Hybrid Fuzzy Controller.

I. INTRODUCTION

The actual energy conversion efficiency of PV module is rather low and is affected by the weather conditions and output load [1-2]. So, to overcome these problems and to get the maximum possible efficiency, the design of all the elements of the PV system has to be optimized. The PV array has a highly nonlinear current-voltage characteristics varying with solar illumination and operating temperature [3-4], that substantially affects the array output power. At particular solar illumination, there is unique operating point of PV array at which its output power is maximum. Therefore, for maximum power generation and extraction efficiency, it is necessary to match the PV generator to the load such that the equilibrium

operating point coincides with the maximum power point of the PV array. The Maximum Power Point Tracking (MPPT) control is therefore critical for the success of the PV systems [5]. In addition, the maximum power operating point varies with insolation level and temperature. Therefore, the tracking control of the maximum power point is a complicated problem. To mitigate these problems, many tracking control strategies have been proposed such as perturb and observe [6], incremental conductance [7], parasitic capacitance [8], constant voltage [9], reactive power control [10]. These strategies have some disadvantages such as high cost, difficulty, complexity and instability. In an effort to overcome aforementioned disadvantages, several

researches have used artificial intelligence such as Fuzzy Logic Controller [11-14] and Artificial Neural Network [15]. Although these methods are effective in dealing with the nonlinear characteristics of the current-voltage curves, they require extensive computation. Optimization algorithms like Genetic Algorithm (GA) and firefly algorithm have attracted the attention in MPPT and controller design. However, these algorithms appear to be effective for the design problem, these algorithms pain from slow convergence in refined search stage, weak local search ability and algorithms may lead to possible entrapment in local minimum solutions. A relatively newer evolutionary computation algorithm, called Artificial Bee Colony (ABC) has been presented by and further established recently by. It is a very simple, robust, and population based stochastic optimization algorithm. In addition, it requires less control parameters to be tuned. Hence, it is suitable optimization tool for locating the maximum power point (MPP) regardless of atmospheric variations.

II. PROPOSED SYSTEM CONFIGURATION

Fig.1 shows a conventional scheme of SPV fed boost converter based water pumping system utilizing SRM drive. It consists of a SPV array, a boost converter, a battery with a bidirectional converter and a three phase SRM together with its drive circuit and water pump which is proposed in [23]. This type of scheme has its own drawbacks like reduction in efficiency and reliability due to the operating voltage of the whole system which is fully regulated by the battery in spite of the SPV array and additional losses due to the bidirectional converter. The other limitations of this system are the absence of MPPT operation, high SRM inverter losses due to PWM switching and its complexity. Another scheme of SPV fed water pumping using SRM drive is proposed in

[26] as shown in Fig.2. It consists of a storage battery connected to the SPV array through a battery voltage regulator (BVR) with the help of two switches. So apart from battery issues, switching losses are also increased. Moreover, addition to these losses, the absence of the MPPT algorithm also decreases the efficiency of the system. The switching losses are minimized by using a concept of variable DC-link voltage for speed control of motor drive [27]. Fig.3 shows the proposed SPV array fed water pumping system utilizing a Cuk converter and employing SRM drive. It consists of the SPV array, a Cuk converter, the mid-point converter feeding the SRM and a coupled water pump. The Cuk converter is so designed and its parameters are selected to operate in CCM mode.

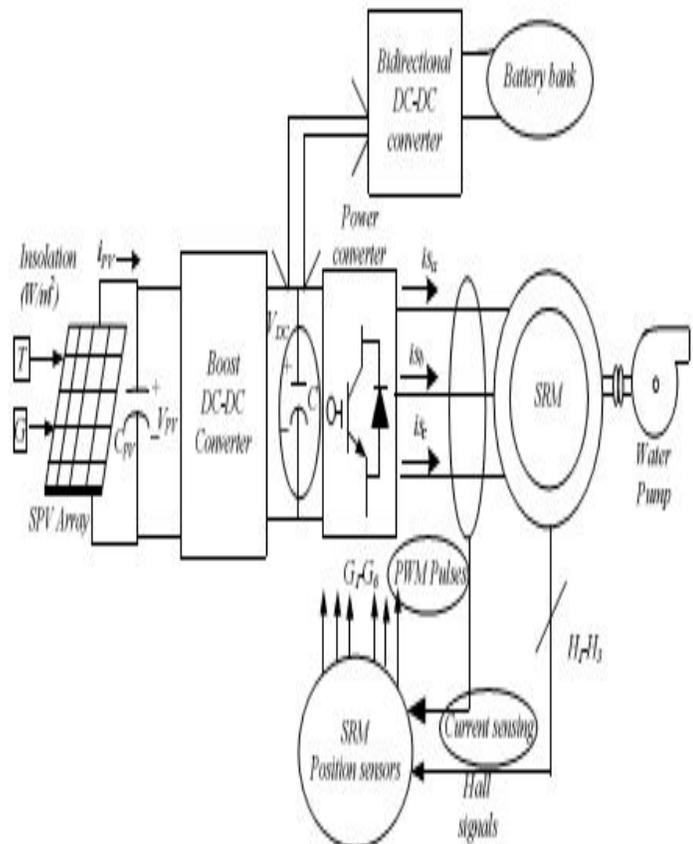


Fig.1. Conventional SPV+ battery fed SRM driven water pumping system [23].

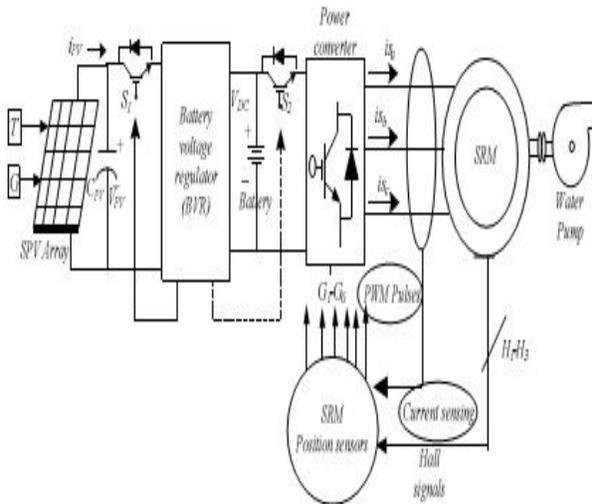


Fig.2. Conventional SPV fed SRM driven water pumping system using BVR[26].

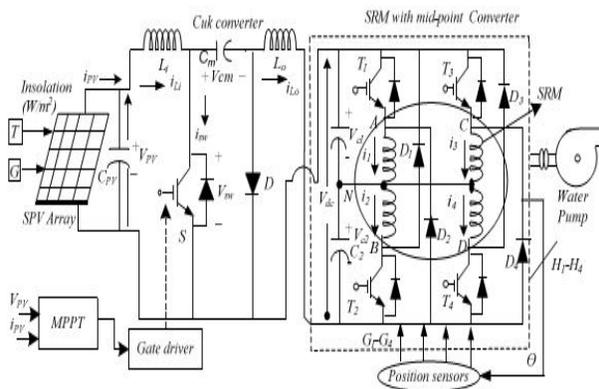


Fig.3. Proposed SPV fed SRM driven water pumping system utilizing Cuk converter

The CCM operation of Cuk converter helps to reduce the current and voltage stresses on its devices and to realize the DC-DC conversion ratio independent of load. The problems present in discontinuous conduction mode (DCM) like ringing phenomenon which is due to the transition of voltage at the other end of the inductor when the current reaches zero, cause EMI noise and increased switch voltage rating which are eliminated in CCM operation of DC-DC Cuk converter. The pulses for fundamental

switching of the midpoint converter switches are generated from the Hall effect position sensors situated on the stator part of SRM and its help to reduce the loss associated with switches of a mid-point converter. The SPV array output power is optimized by Inc MPPT technique.

III. DESIGN AND MODELING OF PROPOSED SYSTEM

The configuration of proposed water pumping system driven by SRM drive is shown in Fig.3. The components of proposed system are designed as per the requirement of SPV fed pump system. A SPV array of 900W and a four phase SRM of 750W power rating are selected. A four switch split capacitor midpoint converter of 320V DC link voltage is selected for proposed system. According to the power rating of the pump and a SRM drive, each stage of the proposed system is designed as follows.

A. Design and Modeling of SPV Array

A SPV array of 900W peak power capacity, somewhat higher than SRM drive rating (750W) is chosen so by considering some losses are always associated with converters and the motor. All the parameters of SPV array are estimated at 1000W/m² insolation level. A solar PV module has short circuit module current (I_{sc}) of 3.55A and open circuit module voltage (V_{ocn}) of 21 V. Each SPV module has a capacity of 50W. The electrical specifications of HB-12100 and designed SPV array at 1000 W/m². Thus, an array of 900W peak power capacity is designed with 1 module in parallel and 17 modules in series with a PV array of 1*17 modules [28].

B. Design and Modeling of Cuk Converter

The DC-DC Cuk converter is designed in such a way that it always operates in CCM regardless of the environmental conditions. The

peak and RMS currents are substantially lower in CCM resulting in lower losses in the conduction paths and smaller ringing because the energy stored in inductances is proportional to the square of the current. The rated DC voltage of the SRM is as, $V_{dc} = 320$ V and the PV voltage at MPP is as, $V_{pv} = V_{mpp} = 289$ V. The relationship between the duty ratio, D of the insulated gate bipolar transistor (IGBT) switch, output voltage, V_{dc} and input voltage, V_{pv} of the Cuk converter is given as [27].

The duty ratio, D of the Cuk converter is estimated as [27],

$$\frac{V_{dc}}{V_{pv}} = -\frac{D}{1-D} \Rightarrow D = \frac{V_{dc}}{V_{dc} + V_{pv}} = \frac{320}{320 + 289} = 0.52 \quad (1)$$

The selected values for DC link capacitors are estimated as [29],

$$C_1 = C_2 = \frac{I(30 - \alpha)}{2\omega\Delta V_{dc}} \quad (2)$$

where, I = DC link current, ω = rated angular speed of SRM, α = conduction angle, ΔV_{dc} = amount of permitted ripple in the voltage across DC link capacitors C_1 & C_2 i.e. 1.5% of V_{dc} . Considering P_{in} as 900 W, V_{dc} as 320 V, f as 50 Hz and $\Delta V_{c1} = \Delta V_{c2}$ as 1.5% of $V_{c1,2}$, the obtained value of ' I ' is 2.34 A and the obtained value of $C_1 = C_2$ is 2441 μ F; hence, C_1 and C_2 are selected as 2500 μ F.

C. Design and Modeling of SRM

The equivalent circuit of SRM is modeled as a current-controlled voltage source as shown in Fig.4. In this equivalent circuit, $e(t)$ is the e.m.f. of the SRM. Due to the saliency on rotor and stator side, SRM has non-sinusoidal current and flux across all four windings against the pulse voltage supply. The modeling is carried out on the supposition that the magnetic coupling between two

consecutive windings of SRM is negligible and its phase inductance profile has the non-linear shape [17]-[18]. Fig.3.5 shows the developed simlink model for 750W, 8/6 pole, 1500rpm SRM.

The expression obtained after applying KVL in conducting phase of SRM is as,

$$v = Ri + \frac{d\psi}{dt} = Ri + \frac{d(Li)}{dt} \quad (3)$$

$$= Ri + L \frac{di}{dt} + i \frac{dL}{dt} \quad (4)$$

$$= Ri + L \frac{di}{dt} + i \frac{dL}{d\theta} * \frac{d\theta}{dt} = Ri + L \frac{di}{dt} + \omega_m i \frac{dL}{d\theta} \quad (5)$$

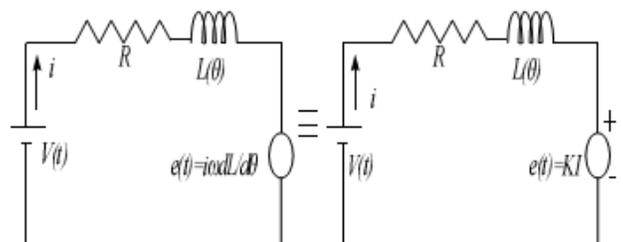


Fig.4 Equivalent circuit of SRM

where, ' V ' is the terminal voltage, i is the current, ψ is the flux linkage in volt-seconds, R is the phase resistance, L is the phase inductance, θ is the rotor position, and ω_m is the angular velocity in rad/s. The last term is sometimes interpreted as a 'back-EMF' as,

$$e = \omega_m i \frac{dL}{d\theta} \quad (6)$$

The instantaneous electrical power $V*i$ as,

$$V_i = Ri^2 + Li \frac{di}{dt} + \omega_m i^2 \frac{dL}{d\theta} \quad (7)$$

The rate of change of magnetic stored energy at any instant is given as,

$$\frac{d}{dt} \left(\frac{1}{2} Li^2 \right) = \frac{1}{2} i^2 \frac{dL}{dt} + Li \frac{di}{dt} = \frac{1}{2} i^2 \omega_m \frac{dL}{d\theta} + Li \frac{di}{dt} \quad (8)$$

The mechanical power conversion is as,

$$P = T\omega_e \quad (9)$$

Thus from (7-8) & (9), the developed torque is as,

$$T = 0.5i^2 dL/d\theta \quad (10)$$

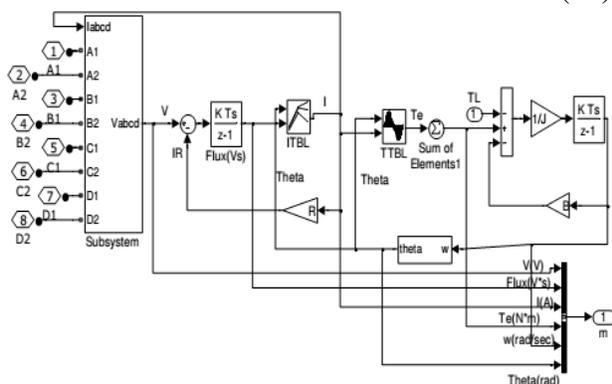


Fig.5 Developed simulink model of SRM

IV. OPERATION AND CONTROL OF PROPOSED SYSTEM

In the proposed system, maximum power of the SPV array is tracked using I_{nc} MPPT technique and the mid-point converter feeding current pulses to the SRM is operated through the electronic commutation.

A. MPPT Control

The location of MPP in the I-V characteristics of SPV array is not predicted in advance and always varies dynamically with insolation levels and environmental conditions. The governing equations which explain the operating principle of I_{nc} method are as [7]-[8],

$$V_{pvref}(k) = V_{pvref}(k-1) + \text{step}, \text{ if } \frac{\Delta I}{\Delta V} > \frac{-I_{PV}}{V_{PV}} \quad (11)$$

$$V_{pvref}(k) = V_{pvref}(k-1) - \text{step}, \text{ if } \frac{\Delta I}{\Delta V} < \frac{-I_{PV}}{V_{PV}} \quad (12)$$

Where, ΔI & ΔV = change in PV current and voltage in two consecutive samples.

The reference voltage of SPV array is checked for upper and lower limits which are set to $0.7V_{ocmax}$ to $0.9V_{ocmax}$. In case the V_{pvref} is between the limits, it is kept as it is else the V_{pvref} is saturated to the nearest limit. The saturation block output is designated as new reference PV voltage (V_{pvrefn}). The V_{pvrefn} and sensed PV voltage are then used to estimate the duty ratio for the Cuk converter. The governing equation for estimating duty ratio is,

$$D(k) = 1 - \frac{V_{pv}}{v_{pv} + v_{pvref}} \quad (13)$$

This reference duty ratio is compared with saw-tooth waveform to generate switching logic for the switch of the Cuk converter.

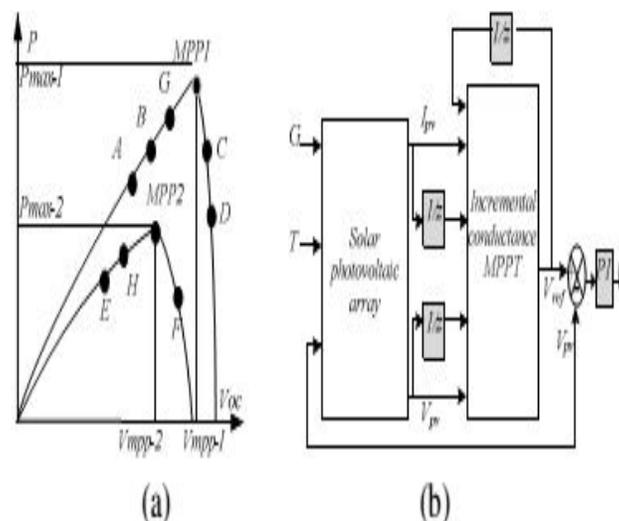


Fig.6. Explanation of I_{nc} method and generation of duty cycle for DC-DC converter by I_{nc} algorithm

Fig.6(a) shows a PV curve along with all possible points on curve undergoing into execution of I_{nc} algorithm. Fig.6(b) explains the method of duty cycle generation through MPPT method and it can be summarized by all possible conditions for MPPT operations using Fig.6 (a).

B. Control of Mid-Point Converter

The complete switching scheme of a mid-point converter is controlled by three parameters; the advance angle (turn-on angle), θ_{on} , turn-off angle, θ_{off} , and value of the effective DC link voltage. The switching angles are defined for each phase based on the rotor position estimation provided by a position. Hall sensors located on the stator of SRM [29]. A mid-point converter injects unipolar current pulses at desired rotor positions to SRM drive. It also controls the magnitude of current for efficient operation of the motor drive and IGBT switches [30]-[32]. The voltage equation at starting is as,

$$1/2 V_{dc} = L (dI/dt) = L (dI/d\theta) (d\theta/dt) \quad (14)$$

$$= L \omega (dI/d\theta) \text{ as } \omega = (d\theta/dt) \quad (15)$$

$$\approx d\theta = 2 L \omega dI/V_{dc} ; \quad (16)$$

Therefore, Turn On angle:

$$\theta_{on} = 2 I L_{min} \omega_{act} / V_{dc} \quad (17)$$

Turn Off angle:

$$\theta_{off} = 2 k I L_{max} \omega_{act} / V_{dc} \quad (18)$$

Where, L_{min} and L_{max} are minimum and maximum inductance values. The winding excitation sequence is decided by the direction of rotation and inductance profile. For forward motion, the sequence is A-B-C-D-A with excitation in the rising inductance region. Fig.3.3 shows the basic circuit diagram of split capacitor mid-point converter. The excitation patterns of all four phases of SRM are symmetrical in nature.

V. HYBRID FUZZY CONTROLLER

The objective of the hybrid controller is to utilize the best attributes of the PI and fuzzy logic controllers to provide a controller which will produce better response

than either the PI or the fuzzy controller. There are two major differences between the tracking ability of the conventional PI controller and the fuzzy logic controller. Both the PI and fuzzy controller produce reasonably good tracking for steady-state or slowly varying operating conditions. However, when there is a step change in any of the operating conditions, such as may occur in the set point or load, the PI controller tends to exhibit some overshoot or oscillations. The fuzzy controller reduces both the overshoot and extent of oscillations under the same operating conditions. Although the fuzzy controller has a slower response by itself, it reduces both the overshoot and extent of oscillations under the same operating conditions. The desire is that, by combining the two controllers, one can get the quick response of the PI controller while eliminating the overshoot possibly associated with it. Switching Control Strategy the switching between the two controllers needs a reliable basis for determining which controller would be more effective.

The answer could be derived by looking at the advantages of each controller. Both controllers yield good responses to steady-state or slowly changing conditions. To take advantage of the rapid response of the PI controller, one needs to keep the system responding under the PI controller for a majority of the time, and use the fuzzy controller only when the system behavior is oscillatory or tends to overshoot. Thus, after designing the best stand-alone PI and fuzzy controllers, one needs to develop a mechanism for switching from the PI to the fuzzy controllers, based on the following two conditions:

- Switch when oscillations are detected;
- Switch when overshoot is detected.

The switching strategy is then simply based on the following conditions: IF the system has an oscillatory behavior then fuzzy controller is activated, Otherwise PI

controller is operated. IF the system has an overshoot then fuzzy controller is activated, Otherwise PI controller is operated. The system under study is considered as having an overshoot when the error is zero and the rate of change in error is any other value than zero. The system is considered oscillatory when the sum of the absolute values of the error taken over time does not equal the absolute values of the sum of the error over the same period of time. Since the system is expected to overshoot during oscillatory behavior, the only switching criterion that needs to be considered is overshoot. However, in practice, it is more convenient to directly implement the control signal according to the control actions delivered by the controller. Consequently, the fuzzy controller can be designed so that normal behavior (no oscillations or overshoot) results in a null fuzzy action as shown in Fig.4. Accordingly, the switching between the two controllers reduces to using PI if the fuzzy has null value; otherwise, the fuzzy output is used. In particular, the fuzzy controller can be designed so that a normal behavior.

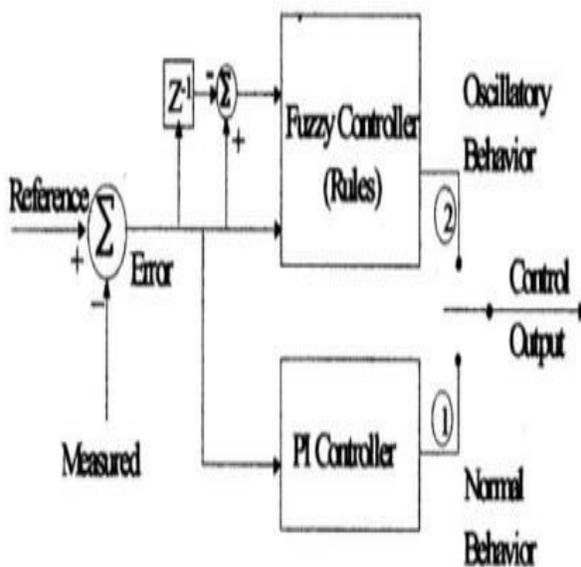


Fig.7 Structure of switching strategy results in a null fuzzy action.

VI. MATLAB/SIMULATION RESULTS

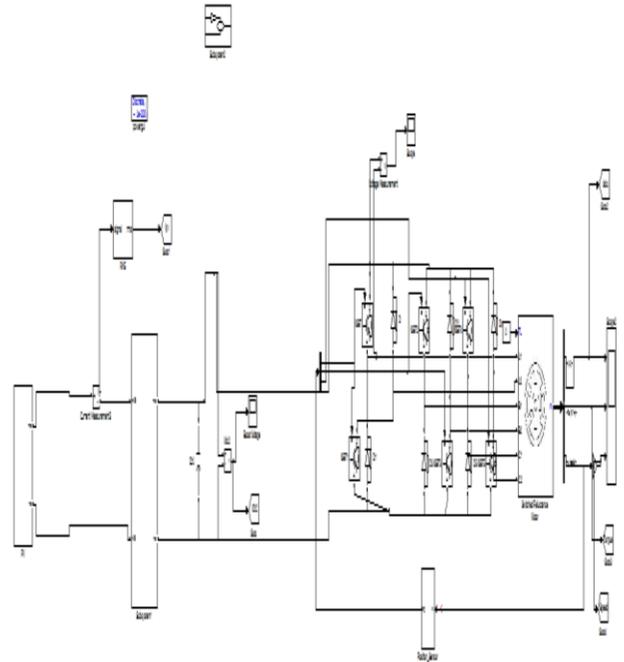


Fig.8 MATLAB/SIMULINK circuit for PV fed Cuk converter with SRM drive

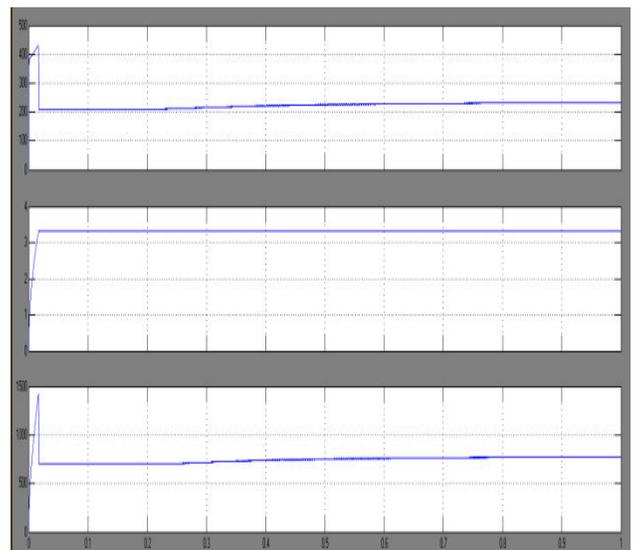


Fig.9 PV voltage, Current and Power

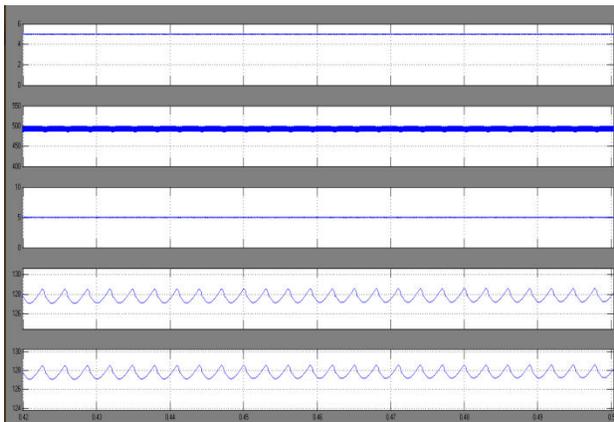


Fig.10 Capacitor voltage 1, Energy transfer capacitor voltage, Capacitor voltage 1, Load current 1 and Load current 2

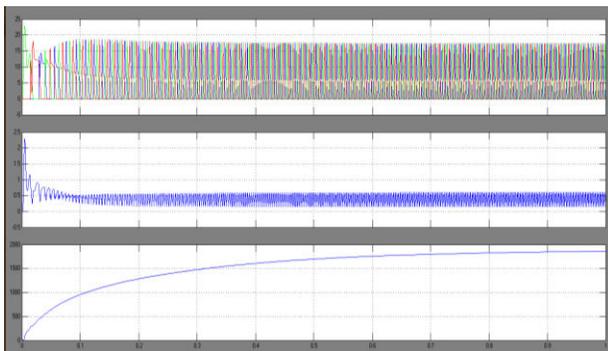


Fig.11 Current, Torque and Speed

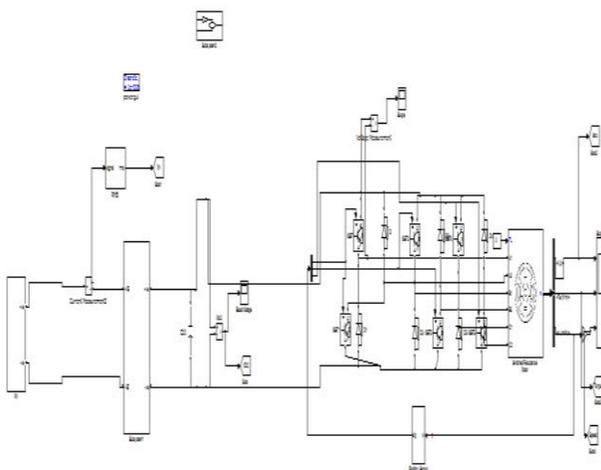


Fig.12 MATLAB/SIMULINK circuit for PV fed Cuk converter with SRM drive

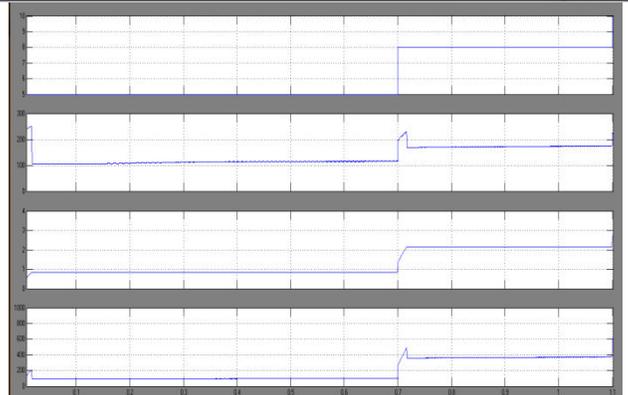


Fig.13 Cuk converter duty cycle, PV voltage, PV current and Power

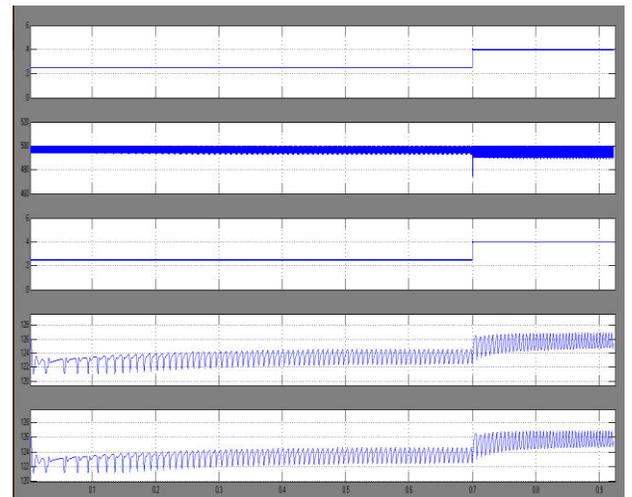


Fig.14 Capacitor voltage 1, Energy transfer capacitor voltage, Capacitor voltage 1, Load current 1 and Load current 2

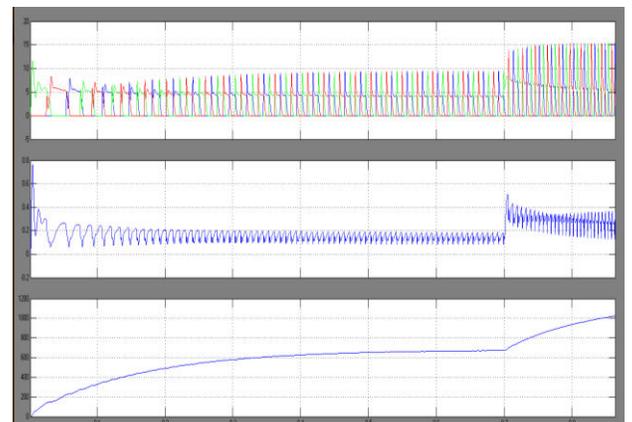


Fig.15 Current, Torque and Speed

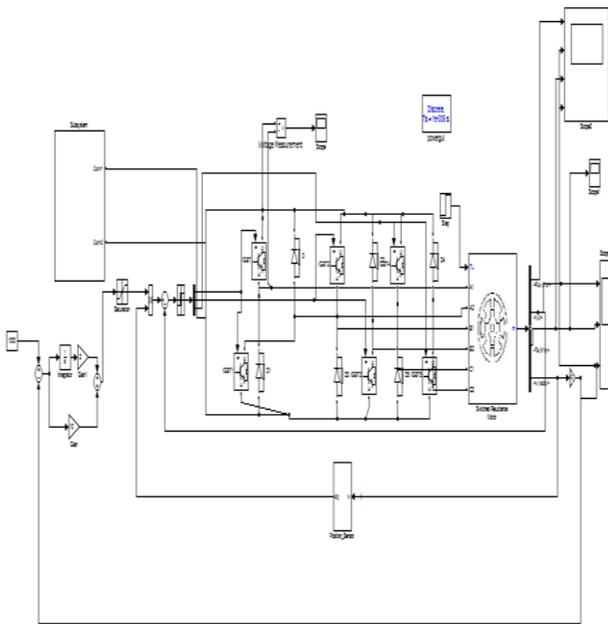


Fig.16 MATLAB/SIMULINK circuit for PV fed Cuk converter with SRM drive with PI controller

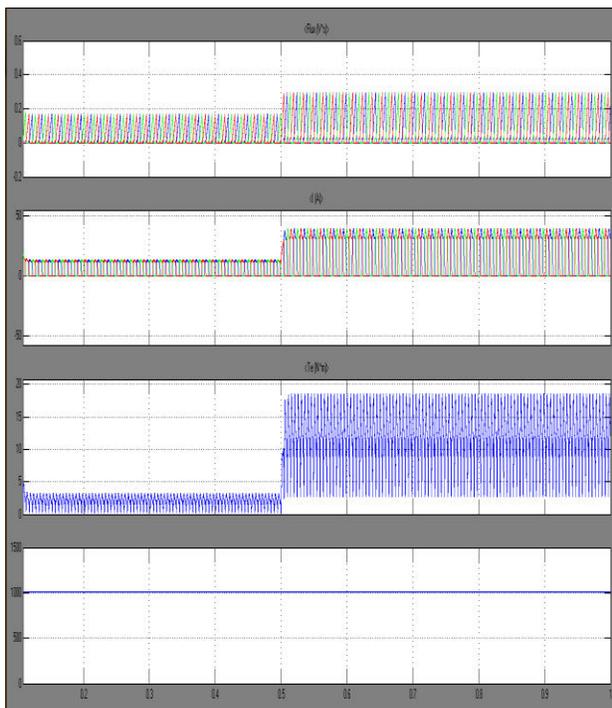


Fig.17 Flux, Current, Torque and Speed

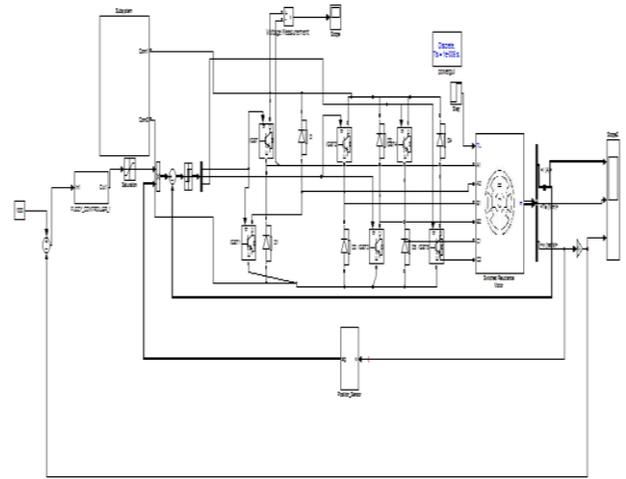


Fig.18 MATLAB/SIMULINK circuit for PV fed Cuk converter with SRM drive with Hybrid Fuzzy Controller

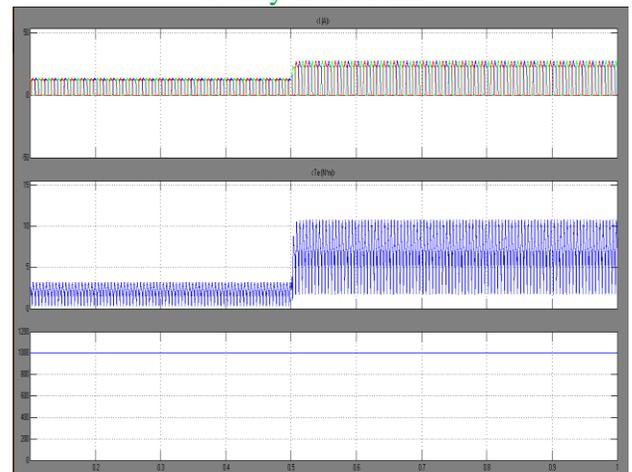


Fig.19 Current, Torque and Speed

VII. CONCLUSION

The SPV array- Cuk converter fed Hybrid fuzzy controller based SRM motor-pump has been proposed and its suitability has been demonstrated through simulated results and experimental validation. The proposed system has been designed and modelled appropriately to accomplish the desired objectives and validated to examine various performances under starting, dynamic and steady state conditions. The performance evaluation has justified the combination of Cuk converter and SRM motor for SPV array based

water pumping. The system under study has shown various desired functions such as MPP extraction of the SPV array, soft starting of SRM motor, fundamental frequency switching of VSI resulting in a reduced switching losses, speed control of SRM motor without any additional control and an elimination of phase current and DC link voltage sensing, resulting in the reduced cost and complexity. The proposed system has operated successfully even under minimum solar irradiance. Finally the hybrid fuzzy control system was also very stable showing a good performance and outperforming all the other classical controllers and the same system conditions.

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BIOGRAPHY:



T. Sravan Kumar is pursuing M .Tech at QIS College of Engineering & Technology (Autonomous) under JNTU, Kakinada. He received his B.Tech degree from RISE Prakasam Group of Institutions JNTU, Kakinada. His current research interests are Power electronics & Power system.



Sri. **R. Sathish** Kumar received the B.Tech. degree in Electrical & Electronics Engineering from Jawaharlal Nehru Technological University, Hyderabad, India, 2005 & M.Tech. degree in Electrical & Electronics Engineering from Jawaharlal Nehru Technological University, Hyderabad, India, in 2010. Currently, he is working as an Associate Professor in QIS College of Engineering and Technology, Ongole, India. He has published a number of papers in various national & international journals & conferences. His research areas are power system operation & control and economic load dispatch.